



Polynomial accessibility condition for the multi-input multi-output nonlinear control system

Ülle Kotta^a, Maris Tõnso^{a*}, and Yu Kawano^b

^a Institute of Cybernetics at Tallinn University of Technology, Akadeemia tee 21, 12618 Tallinn, Estonia; kotta@cc.ioc.ee

^b Graduate School of Informatics, Kyoto University, Sakyo-ku, Kyoto, 606-8501, Japan; ykawano@i.kyoto-u.ac.jp

Received 31 July 2013, revised 18 November 2013, accepted 19 November 2013, available online 20 May 2014

Abstract. The paper presents a computation-oriented necessary and sufficient accessibility condition for the set of nonlinear higher-order input-output differential equations. The condition is presented in terms of the greatest common left divisor of two polynomial matrices, associated with the system of input-output equations. The basic difference from the linear case is that the elements of the polynomial matrices belong to a *non-commutative* polynomial ring. The condition found provides a basis for finding the accessible representation of the set of input-output equations, which is a suitable starting point for the construction of an observable and accessible state space realization. Moreover, the condition allows us to check the transfer equivalence of two nonlinear systems.

Key words: nonlinear control systems, accessibility condition, transfer equivalence, computer algebra system.

1. INTRODUCTION

Controllability is one of the fundamental concepts of mathematical control theory. In the nonlinear case the famous Kalman rank condition branches into many different controllability notions, and among them, the accessibility property stands out as a conceptually tractable and intuitively clear concept [23]. The special notion of controllability, accessibility, refers to the case where every non-constant function of the state is eventually influenced by the control variable of the system and hence cannot satisfy any autonomous differential equation [6].

Our study is motivated by three sources. The first of them is the fundamental problem of accessibility itself. There exist input–output (*i/o*) equations of interest that lack state-space realizations and therefore the techniques that rely on state-space characterization of accessibility are unsuitable for such systems. Second, the accessibility property plays a crucial role in the minimal realization problem; and third, accessibility is related to the concept of transfer equivalence of two systems.

The goal of this paper is to suggest an accessibility definition as well as a computation-oriented necessary and sufficient accessibility condition for nonlinear systems, described by the set of higher-order *i/o* differential equations, not necessarily transformable into the state-space form. Our accessibility definition is based on the concept of autonomous variable, introduced in [21], since this definition is not directly linked to the state and is therefore flexible to be applied to various system classes. Formerly, this concept was called irreducibility of *i/o* equations but it is more natural to call it accessibility since in the case of the realizable *i/o* equations, our definition agrees with that given for the state equations: accessible *i/o* equations have accessible realization. This is demonstrated by Example 2 below.

* Corresponding author, maris@cc.ioc.ee

The results of this paper are based on the conference paper [13]. Compared with [13], in the present paper irreducible system representation (accessible subsystem) is related to the recently introduced concept of transfer matrix of the nonlinear system [8,9], whereas that in [13] was based on the notion of an irreducible differential form, associated with the control system [6]. Moreover, the notion of the transfer equivalence of systems is now defined via the equality of transfer matrices, exactly like in the linear case, and the reduction problem is straightforwardly addressed. Furthermore, the proof of the main theorem is improved (while only the sketch of the proof was given in [13]) and the role of non-uniqueness of the greatest common left divisor to the solution is explained. Finally, comparison with alternative algebraic accessibility criteria is added. Note that the single-input single-output (SISO) case was studied in [27]. For discrete-time nonlinear systems the accessibility property was examined in [16], though the paper itself was focused on finding the minimal realization of the set of i/o equations and accessibility was studied only from the viewpoint of irreducibility of the i/o description of the system. The irreducibility problem for the SISO case has also been treated in [12], being a generalization of [27] to the systems defined on homogeneous time-scales.

The paper is organized as follows. Section 2 describes the differential field and a polynomial matrix representation, associated to nonlinear systems. Using the polynomial matrix description, in Section 3 necessary and sufficient accessibility condition is given. Section 4 is devoted to the system reduction and concept of transfer equivalence. In Section 5 the obtained results are compared with the results of [4] and illustrated by two examples, one realizable in the state-space form and the other not. Section 6 draws the conclusions and drafts some future goals of study.

2. POLYNOMIAL MATRIX DESCRIPTION

Consider a multi-input multi-output (MIMO) nonlinear system described by the set of higher-order i/o differential equations relating the inputs u_k , $k = 1, \dots, m$, the outputs y_i , $i = 1, \dots, p$, and a finite number of their time derivatives:

$$y_i^{(n_i)} = \phi_i(y_j, \dot{y}_j, \dots, y_j^{(n_{ij})}, j = 1, \dots, p, u_k, \dot{u}_k, \dots, u_k^{(r_{ik})}, k = 1, \dots, m), \quad (1)$$

$$i = 1, \dots, p.$$

In (1) ϕ_i are supposed to be real analytic functions. Notations $u := [u_1, \dots, u_m]^T$, $y := [y_1, \dots, y_p]^T$, $n := n_1 + \dots + n_p$, and $r := \max\{r_{ik}\}$ are used below for system (1). Moreover, we assume that the indices in (1) satisfy the relations

$$n_1 \leq n_2 \leq \dots \leq n_p, \quad n_{ij} < n_j, \quad (2)$$

$$n_{ij} < n_i, \quad j \leq i, \quad n_{ij} \leq n_i, \quad j > i,$$

$$r_{ik} < n_i.$$

The conditions (2) mean that Eqs (1) are assumed to be in the form which is an extension of the echelon canonical matrix fraction description, introduced in [22] for linear systems. This form is preferred since it allows the explicit definition of the time-derivative operator in the differential field, associated with the control system. This aspect is important in Mathematica[®] implementation (see below). However, the main results of the paper may also be proved (using a somewhat different mathematical setup) for the implicit system description. Moreover, if the well-defined i/o equations are not in the form (1), one may apply the i/o equivalence transformations¹ from [11,24,25] to bring the system equations into the form (1).

Below we give a brief exposition of the linear algebraic approach, following [6]. Let \mathcal{R} denote the ring of analytic functions in a finite number of (independent) variables from the set $\mathcal{C} = \{y_i, \dots, y_i^{(n_i-1)}, i =$

¹ The transformation is called i/o equivalence transformation if the set of solutions of the original equations is the same as that of the transformed equations.

$1, \dots, p, u_k^{(\beta)}, k = 0, \dots, m, \beta \geq 0\}$. Define the time derivative operator $d/dt : \mathcal{R} \rightarrow \mathcal{R}$, associated with system (1). For that purpose we first define d/dt for the elements of \mathcal{C}

$$\frac{d}{dt} y_i^{(\alpha)} := y_i^{(\alpha+1)}, \quad i = 1, \dots, p, \quad \alpha = 0, \dots, n_i - 2,$$

$$\frac{d}{dt} y_i^{(n_i-1)} := \phi_i(\cdot), \quad i = 1, \dots, p,$$

$$\frac{d}{dt} u_k^{(\beta)} := u_k^{(\beta+1)}, \quad k = 1, \dots, m, \quad \beta \geq 0,$$

and then for $a(\{y_i^{(\alpha)}, u_k^{(\beta)}\}) \in \mathcal{R}$ by

$$\frac{d}{dt} a(\cdot) = \sum_{i=1}^p \sum_{\alpha=0}^{n_i-1} \frac{\partial a}{\partial y_i^{(\alpha)}} \cdot \frac{d}{dt} y_i^{(\alpha)} + \sum_{k=1}^m \sum_{\beta \geq 0} \frac{\partial a}{\partial u_k^{(\beta)}} \cdot \frac{d}{dt} u_k^{(\beta)}.$$

Sometimes the notation $\dot{a} := d/dt(a)$ is also used. Note that whenever $y_i^{(n_i)}$ occurs in the expression, it has to be replaced by $\phi_i(\cdot)$ determined by the i/o equations (1). The pair $(\mathcal{R}, d/dt)$ is a *differential ring* (see [10]). The ring \mathcal{R} is an *integral domain*, i.e. it does not contain any zero divisors. That is, $\forall a, b \in \mathcal{R}, ab = 0 \Rightarrow a = 0$ or $b = 0$. Let $\mathcal{S} := \mathcal{R} \setminus \{0\}$, and consider the set of left fractions denoted as $\mathcal{K} := \mathcal{S}^{-1}\mathcal{R}$. Elements of \mathcal{K} are meromorphic functions in the form $b^{-1}a$, where $a \in \mathcal{R}, b \in \mathcal{S}$. Since \mathcal{R} is an integral domain, the set \mathcal{K} proves to be the field of fractions. The derivative operator d/dt can now be extended so that $d/dt : \mathcal{K} \rightarrow \mathcal{K}$. For $b^{-1}a \in \mathcal{K}$ we define

$$\frac{d}{dt} (b^{-1}a) := (b^2)^{-1}(\dot{a}b - ab\dot{b}), \quad a \in \mathcal{R}, \quad b \in \mathcal{S}.$$

Thus the pair $(\mathcal{K}, d/dt)$ is a *differential field*.

Over the field \mathcal{K} one can define a vector space $\mathcal{E} := \text{span}_{\mathcal{K}}\{d\varphi \mid \varphi \in \mathcal{K}\}$, spanned by the ordinary differentials of the elements of \mathcal{K} . For the definition of the operator $d : \mathcal{K} \rightarrow \mathcal{E}$ see [6]. The paper [7] demonstrates when and how the operator d as used in this paper differs from Kähler differentials. The derivative $\dot{\omega}$ of $\omega = \sum_i \alpha_i d\varphi_i \in \mathcal{E}$, where $\alpha_i, \varphi_i \in \mathcal{K}$, is defined by $\dot{\omega} = \sum_i \dot{\alpha}_i d\varphi_i + \alpha_i d\dot{\varphi}_i$. Note that operators d and d/dt commute, i.e. for $a \in \mathcal{K}$,

$$\frac{d}{dt}(da) = d\left(\frac{d}{dt}a\right) = d\dot{a}. \tag{3}$$

We say that $\omega \in \mathcal{E}$ is an *exact* one-form if $\omega = da$ for some $a \in \mathcal{K}$. A one-form ω for which $d\omega = 0$ is said to be *closed*. Every exact one-form is closed, but the converse holds only locally (see [6]). A subspace is said to be closed or *integrable* if it has a basis which consists only of closed one-forms.

For system (1) a sequence of subspaces $\mathcal{H}_1 \supset \dots \supset \mathcal{H}_{\kappa^*} \supset \mathcal{H}_{\kappa^*+1} = \mathcal{H}_{\kappa^*+2} = \dots =: \mathcal{H}_{\infty}$ of \mathcal{E} is defined by

$$\begin{aligned} \mathcal{H}_1 &= \text{span}_{\mathcal{K}}\{dy_i^{(\alpha)}, du_k^{(\beta)}\}, \\ \mathcal{H}_{\kappa+1} &= \{\omega \in \mathcal{H}_{\kappa} \mid \dot{\omega} \in \mathcal{H}_{\kappa}\}, \quad \kappa \geq 1, \end{aligned} \tag{4}$$

where $j = 1, \dots, p, k = 1, \dots, m, \alpha = 0, \dots, n_j - 1$ and $\beta = 0, \dots, r$.

2.1. Non-commutative ring of polynomials

The differential field \mathcal{K} and the differentiation operator d/dt induce a non-commutative ring of left differential polynomials $\mathcal{K}[s; d/dt]$. A polynomial $p \in \mathcal{K}[s; d/dt]$ can be uniquely written as

$$p = p_m s^m + p_{m-1} s^{m-1} + \dots + p_1 s + p_0, \quad (5)$$

where s is a formal variable and $p_i \in \mathcal{K}$ for $i = 0, \dots, m$. Polynomial $p \neq 0$ iff at least one of the functions p_i is non-zero. If $p_m \neq 0$, then the integer m is called the *degree* of p and denoted by $\deg(p)$. We set additionally $\deg(0) = -\infty$. The addition of the polynomials is defined in the standard way. However, for $a \in \mathcal{K} \subset \mathcal{K}[s; d/dt]$ the multiplication is defined by the commutation rule

$$s \cdot a := as + \dot{a}. \quad (6)$$

It is easy to see that for $s^2 \cdot a = as^2 + 2\dot{a}s + \ddot{a}$, $a \in \mathcal{K}$, and in general, for $n \geq 0$ we obtain $s^n \cdot a = \sum_{i=0}^n \binom{n}{i} a^{(n-i)} s^i$, where $\binom{n}{i}$ is binomial coefficient.

Proposition 1. [18] *The ring $\mathcal{K}[s; d/dt]$ is an integral domain.*

If $p = p_1 p_2$ and $\deg(p_1) > 0$, then p_1 is called a *left divisor* of p . The degree of p satisfies $\deg(p) = \deg(p_1) + \deg(p_2) = d_1 + d_2$, since there are no zero divisors in $\mathcal{K}[s; d/dt]$. To find the left divisor, one can use the *left Euclidean division algorithm* (see [5]). The main idea behind this algorithm is that for given polynomials p_1 and p_2 in the forms (5), with $\deg p_1 > \deg p_2$, there exists a unique *left quotient* polynomial γ_1 and a unique *left remainder* polynomial p_3 such that

$$p_1 = p_2 \gamma_1 + p_3, \quad \deg p_3 < \deg p_2.$$

2.2. Polynomial matrices

We now consider a class of matrices P whose elements are polynomials $p \in \mathcal{K}[s; d/dt]$ of unbounded degree. We write $\mathcal{K}[s; d/dt]^{v \times q}$ for the set of $v \times q$ matrices with entries in $\mathcal{K}[s; d/dt]$. The purpose of this subsection is to show that like in the linear case when the polynomials have real coefficients, the polynomial matrix with entries in $\mathcal{K}[s; d/dt]$ can be transformed by a sequence of elementary column operations to the lower left triangular form. This result allows us to obtain later the accessibility criterion.

Definition 2. *The following three elementary column operations on the polynomial matrix P are defined:*

1. *interchange of columns i and j ,*
2. *multiplication of column i by nonzero scalar in \mathcal{K} ,*
3. *replacement of column i by itself plus any polynomial multiplied by any other column j .*

Definition 3. *A matrix $U \in \mathcal{K}[s; d/dt]^{v \times v}$ is called unimodular if it is invertible in $\mathcal{K}[s; d/dt]^{v \times v}$, that is, there exists a matrix $U^{-1} \in \mathcal{K}[s; d/dt]^{v \times v}$ such that $U U^{-1} = U^{-1} U = I_v$.*

Any sequence of elementary column operations on P is equivalent to post multiplication (right multiplication) of P by an appropriate unimodular matrix U_R .²

Definition 4. *Two polynomial matrices P and \hat{P} will be called column equivalent iff one of them can be obtained from the other by a sequence of elementary column operations.*

The matrix P is thus column equivalent to \hat{P} if and only if $P = \hat{P} U_R$ where U_R is a unimodular matrix.

² Similarly, any sequence of elementary row operations on P is equivalent to premultiplication (left multiplication) of P by an appropriate unimodular matrix U_L .

Theorem 5. Any $q \times v$ ($q \leq v$) polynomial matrix P is column equivalent to the lower left triangular matrix shown below, i.e. one can always find a sequence of elementary column operations which reduces P to the lower left triangular form where $\tilde{P} = [G_L \ 0]$ and

$$G_L = \begin{pmatrix} g_{11} & 0 & \dots & 0 \\ g_{21} & g_{22} & \dots & 0 \\ \vdots & \vdots & & \vdots \\ g_{q1} & g_{q2} & \dots & g_{qq} \end{pmatrix}. \tag{7}$$

Furthermore, in the above form, the polynomials $g_{k1}, \dots, g_{k,k-1}$ are of lower degree than g_{kk} for all $k = 1, \dots, v$ if $\deg g_{kk} > 0$, and are all zero if g_{kk} is a nonzero scalar in \mathcal{K} .

Proof. If the first row of P , $p_{1i} \neq 0$, one may choose a polynomial of the least degree from its elements and, by permutation of the columns, make it the new $(1, 1)$ entry \tilde{p}_{11} . We then apply the left Euclidean division algorithm to every other nonzero entry in the first row. That is, we divide every other nonzero element \tilde{p}_{1i} in the first row by \tilde{p}_{11} obtaining the quotients \tilde{q}_{1i} and remainders \tilde{r}_{1i} according to the relationship $\tilde{p}_{1i} = \tilde{p}_{11}\tilde{q}_{1i} + \tilde{r}_{1i}$ where either $\tilde{q}_{1i} = 0$ or $\deg \tilde{r}_{1i} < \deg \tilde{p}_{11}$. We then subtract from each nonzero i th column the first column multiplied by \tilde{q}_{1i} . If not all of the remainders \tilde{r}_{1i} are zero, we choose one of the least degree and make it the new $(1, 1)$ entry by another permutation of the columns. The purpose of repeating this process as many times as necessary is to reduce step by step the degree of the polynomial element in the $(1, 1)$ entry. Since the degree of the $(1, 1)$ entry is finite, this repeated process ends at some steps, in particular, when all of the remaining elements of the first row are identically zero.

Next consider the second row of this matrix and, ignoring the first column for the moment, apply the above procedure to the elements beginning with the second column and second row. In this way we zero all the elements to the right of the $(2, 2)$ entry.

If the $(2, 1)$ element is of equal or higher degree than that of the $(2, 2)$ element, the division algorithm can be employed to reduce the $(2, 1)$ element to the remainder term associated with the division of the $(2, 1)$ entry by the $(2, 2)$ entry, or to zero if both elements are scalars. Continuing in this manner, with the elements beginning with the third column and third row next, we eventually reduce P to the appropriate form. \square

Definition 6. If three polynomial matrices satisfy the relation $P = C_L Q$, then C_L is called a left divisor of P and P is called a right multiple of C_L . The greatest common left divisor (gclid) of two polynomial matrices V and W is a common left divisor which is a right multiple of every common left divisor of V and W .

Theorem 7. Consider the pair $\{P, Q\}$ of polynomial matrices which have the same number of rows. If the composite matrix $[P \mid Q]$ is reduced to the lower left triangular form $[G_L \mid 0]$ as in the proof of Theorem 5, then G_L is a gclid of P and Q .

The proof is analogous with the discrete time case (see [16], Theorem 10) and therefore omitted.

The gclid is, in general, not unique, since any two gclids, G_L^1 and G_L^2 , are related by definition as $G_L^1 = G_L^2 W_1$, $G_L^2 = G_L^1 W_2$, where W_1 and W_2 are polynomial matrices and if G_L^1 is nonsingular, then W_1 and W_2 are unimodular. The concept of a gclid now enables us to extend to the matrix case the notion of a pair of relatively left prime polynomials.

Definition 8. A pair $\{P, Q\}$ of polynomial matrices which have the same number of rows is said to be relatively left prime iff their gclids are unimodular matrices.

The notion of a pair of relatively left prime polynomial matrices implies the inability to factor some non-unimodular matrix from the left side of both members of the pair.

2.3. Polynomials as the operators

A left differential polynomial $a \in \mathcal{K}[s; d/dt]$ may be interpreted as an operator $a(s) : \mathcal{E} \rightarrow \mathcal{E}$. Define for $j = 1, \dots, p$, $k = 1, \dots, m$, and $dy_j, du_k \in \mathcal{E}$

$$sdy_j := d\dot{y}_j, \quad sdu_k := d\dot{u}_k. \tag{8}$$

It is natural to extend (8) for $a = \sum_{i=0}^k a_i s^i$ as $a(s)(\alpha d\zeta) := \sum_{i=0}^k a_i (s^i \cdot \alpha) d\zeta$ with $a_i, \alpha \in \mathcal{K}$ and $d\zeta \in \{dy_1, \dots, dy_p, du_1, \dots, du_m\}$. Using (8), every one-form

$$\omega = \sum_{j=1}^p \sum_{\alpha=0}^{n_j-1} a_{j,\alpha} dy_j^{(\alpha)} + \sum_{k=1}^m \sum_{\beta=0}^{\ell} b_{k,\beta} du_k^{(\beta)} \in \mathcal{E},$$

where $a_{j,\alpha}, b_{k,\beta} \in \mathcal{K}$ and $\ell \geq 0$, may be expressed in terms of the left differential polynomials as

$$\omega = \sum_{j=1}^p \left(\sum_{\alpha=0}^{n_j-1} a_{j,\alpha} s^\alpha \right) dy_j + \sum_{k=1}^m \left(\sum_{\beta=0}^{\ell} b_{k,\beta} s^\beta \right) du_k := [a_1(s), \dots, a_p(s)] dy + [b_1(s), \dots, b_m(s)] du,$$

where $a_j = \sum_{\alpha=0}^{n_j-1} a_{j,\alpha} s^\alpha$, $b_k = \sum_{\beta=0}^{\ell} b_{k,\beta} s^\beta$, $dy = [dy_1, \dots, dy_p]^T$, and $du = [du_1, \dots, du_m]^T$. It is easy to see that $s\omega = \dot{\omega}$, for $\omega \in \mathcal{E}$.

2.4. Polynomial description of linearized equations

By applying the operator d to (1) we obtain for $i = 1, \dots, p$

$$dy_i^{(n_i)} - \sum_{j=1}^p \sum_{\alpha=0}^{n_j} \frac{\partial \phi_i}{\partial y_j^{(\alpha)}} dy_j^{(\alpha)} - \sum_{k=1}^m \sum_{\beta=0}^{r_{ik}} \frac{\partial \phi_i}{\partial u_k^{(\beta)}} du_k^{(\beta)} = 0. \quad (9)$$

Since $dy_j^{(\alpha)} = s^\alpha dy_j$ and $du_k^{(\beta)} = s^\beta du_k$, (9) can be rewritten as

$$P(s)dy + Q(s)du = 0, \quad (10)$$

where $P \in \mathcal{K}[s; d/dt]^{p \times p}$ and $Q \in \mathcal{K}[s; d/dt]^{p \times m}$ are polynomial matrices, whose elements $p_{ij}, q_{ik} \in \mathcal{K}[s; d/dt]$ are

$$p_{ij} = \begin{cases} s^{n_i} - \sum_{\alpha=0}^{n_j} p_{ij,\alpha} s^\alpha, & \text{if } i = j, \\ - \sum_{\alpha=0}^{n_j} p_{ij,\alpha} s^\alpha, & \text{if } i \neq j, \end{cases} \quad q_{ik} = - \sum_{\beta=0}^{r_{ik}} q_{ik,\beta} s^\beta, \quad (11)$$

whereas

$$p_{ij,\alpha} = \frac{\partial \phi_i}{\partial y_j^{(\alpha)}} \in \mathcal{K}, \quad q_{ik,\beta} = \frac{\partial \phi_i}{\partial u_k^{(\beta)}} \in \mathcal{K}.$$

Equation (10) describes the (globally) linearized system, associated with Eqs (1).

3. ACCESSIBILITY OF THE i/o EQUATIONS

Definition 9. [6] A non-constant (possibly vector) function³ φ_r (with components) in \mathcal{K} is said to be an autonomous variable for system (1) if there exist an integer $\mu \geq 1$ and a non-zero meromorphic (again possibly vector) function F such that

$$F(\varphi_r, \varphi_r^{(1)}, \dots, \varphi_r^{(\mu)}) = 0. \quad (12)$$

³ In the notation φ_r the subscript r is abbreviation from the word *reduced*, not a varying index.

Note that φ_r denotes a variable as well as a function of y, u , and their time derivatives. While φ_r is a function of $y^{(i)}$ and $u^{(j)}$, $0 \leq i \leq n - \mu$, $0 \leq j \leq n - \mu$, it is also governed by the *autonomous differential equation* (12). For any initial condition, the solution φ_r is uniquely determined by this autonomous differential equation and is consequently independent of the external input u . In this sense φ_r is an autonomous variable which represents the lack of controllability of the nonlinear system [21].

The notion of autonomous variable can be used to define the accessibility of the nonlinear system (1) as follows:

Definition 10. System (1) is said to be accessible if it does not admit any non-constant autonomous variable. Otherwise the system is called non-accessible.

Theorem 11. The nonlinear system (1) is accessible iff the polynomial matrices P and Q in (10) are relatively left prime.

Proof. Sufficiency. The proof is by contradiction. Suppose that P and Q are relatively left prime but contrary to our claim, system (1) is not accessible. Then, according to Definition 10, there exist (at least one) function $\varphi_r \in \mathcal{K}$, an integer $\mu \geq 1$, and $F = (F_1, \dots, F_p)^T$, such that (12) holds. Denote $\tilde{\varphi} := F(\varphi_r, \dot{\varphi}_r, \dots, \varphi_r^{(\mu)})$. Then $\tilde{\varphi} \in \mathcal{K}$,

$$d\varphi_r = \sum_{j=1}^p \sum_{i=0}^{\ell} \frac{\partial \varphi_r}{\partial y_j^{(i)}} dy_j^{(i)} + \sum_{k=1}^m \sum_{j=0}^{\ell} \frac{\partial \varphi_r}{\partial u_k^{(j)}} du_k^{(j)}, \tag{13}$$

and

$$d\tilde{\varphi} = \sum_{k=0}^{\mu} \frac{\partial F}{\partial \varphi_r^{(k)}} d\varphi_r^{(k)}, \tag{14}$$

where $\ell \leq n - \mu$. Since $s^i dy_j = dy_j^{(i)}$, $s^j du_k = du_k^{(j)}$, and $s^k d\varphi_r = d\varphi_r^{(k)}$, (13) and (14) can be rewritten in terms of left differential polynomials:

$$\begin{aligned} d\varphi_r &= \begin{pmatrix} \sum_{i=0}^{\ell} \frac{\partial \varphi_r}{\partial y_1^{(i)}} s^i & \dots & \sum_{i=0}^{\ell} \frac{\partial \varphi_r}{\partial y_p^{(i)}} s^i \end{pmatrix} \cdot \begin{pmatrix} dy_1 \\ \vdots \\ dy_p \end{pmatrix} + \begin{pmatrix} \sum_{j=0}^{\ell} \frac{\partial \varphi_r}{\partial u_1^{(j)}} s^j & \dots & \sum_{j=0}^{\ell} \frac{\partial \varphi_r}{\partial u_m^{(j)}} s^j \end{pmatrix} \cdot \begin{pmatrix} du_1 \\ \vdots \\ du_m \end{pmatrix} \\ &:= \hat{P}(s)dy + \hat{Q}(s)du \end{aligned}$$

and

$$d\tilde{\varphi} = \sum_{k=0}^{\mu} \frac{\partial F}{\partial \varphi_r^{(k)}} s^k d\varphi_r := \hat{G}(s)d\varphi_r.$$

Then the equation $d\tilde{\varphi} = 0$ can be rewritten as $\hat{G}(s)[\hat{P}(s)dy + \hat{Q}(s)du] = 0$. The remaining part of the proof relies on the fact that $p + v$ differential forms, defined by p rows of $P(s)dy - Q(s)du$ and v rows of $d\tilde{\varphi}$, are dependent. In order to simplify the proof, we assume that F contains a single element, i.e. $F = (F_1)$ and $v = 1$. The proof of the general case is analogous. Let $d\varphi_i := \sum_j p_{ij}(s)dy_j - \sum_k q_{ik}(s)du_k$, $i = 1, \dots, p$, where p_{ij} and q_{ik} are the elements in the i th rows of the matrices P and Q , respectively. Then there exist $\alpha_i \in \mathcal{K}$, $i = 1, \dots, p$, at least one of them non-zero such that $d\tilde{\varphi} = \sum_{i=1}^p \alpha_i d\varphi_i$. Without loss of generality one can assume that $\alpha_1 \neq 0$. Then we have

$$\begin{pmatrix} \hat{G}(s)d\varphi_r \\ d\varphi_2 \\ \vdots \\ d\varphi_p \end{pmatrix} = \begin{pmatrix} \alpha_1 & \alpha_2 & \dots & \alpha_p \\ 0 & 1 & \dots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & \dots & 1 \end{pmatrix} \cdot \begin{pmatrix} d\varphi_1 \\ d\varphi_2 \\ \vdots \\ d\varphi_p \end{pmatrix} \tag{15}$$

yielding

$$\begin{pmatrix} d\varphi_1 \\ d\varphi_2 \\ \vdots \\ d\varphi_p \end{pmatrix} = \begin{pmatrix} \frac{1}{\alpha_1} \hat{G}(s) & -\frac{\alpha_2}{\alpha_1} & \cdots & -\frac{\alpha_p}{\alpha_1} \\ 0 & 1 & \cdots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & \cdots & 1 \end{pmatrix} \cdot \begin{pmatrix} d\varphi_r \\ d\varphi_2 \\ \vdots \\ d\varphi_p \end{pmatrix}.$$

Therefore $P(s)dy + Q(s)du = G_L(s) \cdot [\tilde{P}(s)dy + \tilde{Q}(s)du]$, where

$$G_L = \begin{pmatrix} \frac{1}{\alpha_1} \hat{G}(s) & -\frac{\alpha_2}{\alpha_1} & \cdots & -\frac{\alpha_p}{\alpha_1} \\ 0 & 1 & \cdots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & \cdots & 1 \end{pmatrix}.$$

Since $\deg \hat{G} = \mu > 0$, the gcd of the matrices P and Q , i.e. the matrix G_L is not unimodular. Hence P and Q are not relatively left prime.

Necessity. The proof of the necessity part is by contradiction. Assume that system (1) is accessible but the matrices P and Q are not relatively prime. The latter means that P and Q have a gcd G_L in the form (7), which is not unimodular, such that Eq. (10) can be written as

$$P(s)dy + Q(s)du = G_L(s)[\tilde{P}(s)dy + \tilde{Q}(s)du] = G_L(s)\omega = 0, \quad (16)$$

where

$$\omega = [\omega_1, \dots, \omega_p]^T := \tilde{P}(s)dy + \tilde{Q}(s)du \quad (17)$$

is a column-vector of irreducible differential one-forms.

We proceed to show that the components of one-form ω are exact or can be made exact by multiplying ω with a unimodular matrix U from the left. The elements of ω may be classified as follows: they are either differentials of the original irreducible equations (integrable by definition) or the elements of \mathcal{H}_∞ , defined by (4). According to [1], the subspace \mathcal{H}_∞ is closed. This means that its basis vectors are exact or can be made exact by multiplying ω with a certain unimodular matrix U from the left. This just corresponds to changing the gcd G_L by $G_L U^{-1} =: G'_L$, note that G_L is unique up to multiplication with the unimodular matrix from the right. From (16) we now obtain

$$G_L(s)U^{-1}U\omega = G'_L(s)U^{-1}d\varphi_r = G'_L(s)d\varphi_r = 0,$$

where $d\varphi_r = [d\varphi_{r1}, \dots, d\varphi_{rp}]^T$. The vector of one-forms $G'_L(s)d\varphi_r = P(s)dy + Q(s)du$ corresponds to the original linearized equations (10) and is thus integrable. This means that the coefficients of polynomials in G'_L either depend only on the components of φ_r or are real numbers. Since G'_L is non-unimodular, there exists at least one component of F , for instance F_i , and at least one component of φ_r , for instance φ_{rj} , such that $F_i(\dots, \varphi_{rj}^{(\mu)}, \dots) = 0$, $\mu \geq 1$. According to Definition 9, the system admits autonomous variable φ_{rj} , and hence (1) is not accessible. \square

Remark 12. Observe that unimodular G_L will give $\mu = 0$ in (12), whenever the equations $F_1(\cdot) = 0, \dots, F_p(\cdot) = 0$ are transformed into the form (1); see also the following example.

Example 1. Let $d\varphi_{r1} = d(\dot{y}_1 - u_1)$, $d\varphi_{r2} = d(\dot{y}_2 - u_2 y_1)$, and

$$G_L = \begin{pmatrix} \dot{y}_1 - u_1 & -s \\ -\dot{y}_1 + u_1 & s + 1 \end{pmatrix}$$

be unimodular but involving polynomials of degree 1. Then $G_L[d\varphi_{r1} \ d\varphi_{r2}]^T$ yields, after integration of the one-forms,

$$\begin{aligned}\psi_1 &:= -\dot{y}_2 + u_2 y_1 + u_2 \dot{y}_1 + \frac{1}{2} \dot{y}_1^2 - \dot{y}_1 u_1 + \frac{1}{2} u_1^2 = 0, \\ \psi_2 &:= \dot{y}_2 - u_2 y_1 - u_2 \dot{y}_1 + \dot{y}_2 - u_2 y_1 - \frac{1}{2} \dot{y}_1^2 + \dot{y}_1 u_1 + \frac{1}{2} u_1^2 = 0.\end{aligned}\quad (18)$$

Obviously $\psi_1 = F_1(\cdot) = -\dot{\varphi}_{r2} + \frac{1}{2} \varphi_{r1}^2 = 0$, $\psi_2 = F_2(\cdot) = \dot{\varphi}_{r2} + \varphi_{r2} - \frac{1}{2} \varphi_{r1}^2 = 0$, which creates an illusion as if $\mu = 1$ in F and system (18) is reducible. Next, transform Eqs (18) into the form (1). It can be done following the algorithm in [11], but for this simple example it is easy to see that $\psi_1 + \psi_2 = \dot{y}_2 - u_2 y_1 = 0$ and $\psi_1 + \psi_1 + \psi_2 = \frac{1}{2}(\dot{y}_1 - u_1)^2 = 0$. The obtained equations

$$\begin{aligned}\psi'_1 &:= \dot{y}_1 - u_1 = 0, \\ \psi'_2 &:= \dot{y}_2 - u_2 y_1 = 0\end{aligned}\quad (19)$$

are in the form (1), and now $F'_1 = \varphi_{r1}$, $F'_2 = \varphi_{r2}$, thus $\mu = 0$. Finally, note that by the results of [11], unimodular transformation $d\varphi = U(s)d\varphi_r$ means that systems $\varphi = 0$ ja $\varphi_r = 0$ are i/o equivalent.

4. SYSTEM REDUCTION AND TRANSFER EQUIVALENCE

Note that the gcd of polynomial matrices P and Q is not unique. If G_L is a gcd of P and Q , then multiplying it by a unimodular matrix from the right, we get another gcd of P and Q . In general, we get the class of matrices (gcds) that are column equivalent, i.e. one of them can be obtained from the other by a sequence of elementary column operations. Therefore, if a gcd is unimodular, then all gcds must be unimodular. Taking from the class of column equivalent matrices different gcds, we get different vectors $\omega = \tilde{P}(s)dy - \tilde{Q}(s)du$ that are, however, i/o equivalent [11].

By the proof of Theorem 11 the *irreducible differential forms* ω are either exact or can be made exact by multiplying ω from the left by a unimodular matrix U . So, one can write $\bar{P}(s)dy - \bar{Q}(s)du = d\varphi_r$, where $d\varphi_r = [d\varphi_{r1}, \dots, d\varphi_{rp}]^T$, and $\bar{P} = U\tilde{P}$, $\bar{Q} = U\tilde{Q}$. Then the equations $\varphi_r(\cdot) = 0$ are called *irreducible equations* of system (1).

System reduction. If G_L is the non-unimodular gcd of P and Q , system (1) is not accessible. In such a case one may find the accessible (irreducible) subsystem $\varphi_r(\cdot) = 0$. The matrices \bar{P} and \bar{Q} , describing the irreducible system, are relatively left prime and their gcds are unimodular matrices. If G_L is non-unimodular, then the procedure of finding the new system $\varphi_r = 0$ is called *system reduction*. If G_L is unimodular, then system (1) is already in irreducible form, the new system $\varphi_r = 0$ is *i/o equivalent* with (1) (see [11]), and no reduction occurs. System reduction can be performed by the following algorithm:

Algorithm

Input: Set of i/o equations in the form (1)

Output: Set of reduced i/o equations $\varphi_r = 0$

1. Compute matrices P and Q as in (10)
2. Find $G_L := \text{gcd}(P, Q)$
3. **If** G_L is unimodular **Then** 'System is irreducible' **Else**
 - (a) Solve the equations $P = G_L \tilde{P}$ and $Q = G_L \tilde{Q}$ for \tilde{P} and \tilde{Q}
 - (b) $\omega := \tilde{P}dy + \tilde{Q}du$
 - (c) Integrate ω i.e. find unimodular matrix U and φ_r such that $U\omega = d\varphi_r$
 - (d) Return irreducible equations $\varphi_r = 0$

Transfer equivalence. From the ring $\mathcal{K}[s; d/dt]$ one can construct a non-commutative field of fractions, which is required to describe the transfer matrix of the nonlinear system. Define a set $\mathcal{V} := \mathcal{K}[s; d/dt] \setminus \{0\}$. Consider the set of left fractions denoted by $\mathcal{K}(s; d/dt) = \mathcal{V}^{-1} \mathcal{K}[s; d/dt]$. Elements of $\mathcal{K}(s; d/dt)$ are

left fractions in the form $b^{-1}a$, where $a \in \mathcal{K}[s; d/dt]$, $b \in \mathcal{V}$. Since the ring $\mathcal{K}[s; d/dt]$ is an integral domain, $\mathcal{K}(s; d/dt)$ is the field. Then one can consider a class of matrices, whose elements are left fractions $f \in \mathcal{K}(s; d/dt)$ (see [8]). Let $\mathcal{K}(s; d/dt)^{p \times m}$ be the set of $p \times m$ matrices with their entries from $\mathcal{K}(s; d/dt)$. Multiplying the globally linearized system equations (10) from the left by P^{-1} allows us to rewrite (10) as

$$dy = P^{-1}(s)Q(s)du,$$

whenever the i/o equations (1) are independent, i.e. whenever the Dieudonné determinant of the matrix P is nonzero [14]. The inverse matrix $P^{-1} \in \mathcal{K}(s; d/dt)^{p \times p}$ can be computed by the Gauss–Jordan elimination method, adapted for non-commutative polynomials in [20]. If the polynomial matrices P and Q are not relatively left prime, i.e. $P = G_L \cdot \tilde{P}$ and $Q = G_L \cdot \tilde{Q}$ (G_L is the non-unimodular gcd of P and Q), we get

$$dy = P^{-1}(s)Q(s)du = \tilde{P}^{-1}(s)G_L^{-1}(s)G_L(s)\tilde{Q}(s)du = \tilde{P}^{-1}(s)\tilde{Q}(s)du.$$

Therefore, system (1) can be characterized by a matrix from $\mathcal{K}(s; d/dt)^{p \times m}$. In our case it is the matrix $F = P^{-1}Q = \tilde{P}^{-1}\tilde{Q}$. Note that every matrix of left fractions from $\mathcal{K}(s; d/dt)^{p \times m}$, represented in the form $A^{-1}B$, does not correspond to a control system, since the set of one-forms $A(s)dy + B(s)du$ is not necessarily integrable.

Definition 13. [8] A matrix $F \in \mathcal{K}(s; d/dt)^{p \times m}$ such that $dy = F(s)du$ is said to be the transfer matrix of the MIMO nonlinear system (1).

Definition 14. Two systems Σ_1 and Σ_2 , defined by the equations of the form (1), are called transfer equivalent if they have the same transfer matrix.

From the above definition we can conclude that the irreducible set of equations is transfer equivalent to the original system description (1).

5. DISCUSSION, EXAMPLES, AND MATHEMATICA[®] IMPLEMENTATION

In this section we first compare the notion of accessibility, introduced in this paper, with the notion of controllability, as it is defined in [4] for the set of i/o equations. The comparison obviously holds only for the subclass of systems studied in [4], that is for systems, defined by polynomial equations in variables y , \dot{y} , \ddot{y} , etc. over the field of rational functions of the variables u , \dot{u} , \ddot{u} , etc. For this class of systems the two notions practically coincide. Really, controllability is defined as \mathbb{R} being differentially algebraically closed in the differential field \mathcal{K} , defined by the system equations (1). This just means that system (1) does not admit a variable $\phi_t \in \mathcal{K}$ that satisfies an autonomous differential algebraic equation $F(\phi_t, \dot{\phi}_t, \dots, \phi_t^{(\mu)}) = 0$ for some $\mu \geq 1$. Note that in our paper F may be a meromorphic function. Moreover, the paper [4] does not provide a method for finding the accessible (irreducible) system description, though the method for checking accessibility is given in [17].

The algorithms for checking the accessibility and reduction of a non-accessible (reducible) system have been implemented in the computer algebra system Mathematica[®] as a part of the package NLControl, devoted to modelling, analysis, and synthesis problems of nonlinear control systems. The functions are also available on the NLControl website www.nlcontrol.ioc.ee. The main benefit of the website is that the user does not need Mathematica[®] to be installed into a local computer; only internet connection and a browser are necessary to run the functions. Note the difference in terminology: on the NLControl website the accessibility of the i/o system can be checked using the function **Irreducibility**. The irreducible equations, which also represent the accessible subsystem, can be found by the function **Reduction**.

Example 2. Consider the system described by the i/o differential equations

$$\begin{aligned} \ddot{y}_1 &= -y_2 + u_2 y_2 + y_2 \dot{u}_1 + y_2 \ddot{u}_2 - \dot{y}_1 - \dot{y}_2 + u_1 \dot{y}_2 + u_2 \ddot{y}_2 - \ddot{y}_2, \\ \ddot{y}_2 &= -3y_2 + 2u_1 y_2 + 3u_2 y_2 - 3\dot{y}_1 - 2\dot{y}_2 + 2\dot{u}_1 \dot{y}_2 + y_2 \ddot{u}_1 + u_1 \ddot{y}_2. \end{aligned} \quad (20)$$

System (20) satisfies conditions (2). Linearized equations of (20) as in (10) are determined by the polynomial matrices

$$P = \begin{pmatrix} s^2 + s & s^2 + (1 - u_1 - u_2)s + 1 - u_2 - \dot{u}_1 - \dot{u}_2 \\ -3s & s^3 - u_1s^2 + (2 - 2\dot{u}_1)s + 3 - 2u_1 - 3u_2 - \ddot{u}_1 \end{pmatrix},$$

$$Q = \begin{pmatrix} -y_2s - \dot{y}_2 & -y_2s - y_2 - \dot{y}_2 \\ -y_2s^2 - 2\dot{y}_2s - 2y_2 - \ddot{y}_2 & -3y_2 \end{pmatrix}.$$

According to Theorem 7, one can find a gcd G_L of P and Q by reducing the composite matrix $[P|Q]$ into the lower left triangular form $[G_L|0]$. Following the algorithm in the proof of Theorem 5 yields

$$G_L = \begin{pmatrix} & y_2 & 0 \\ -y_2s^2 - 2\dot{y}_2s + y_2 - \ddot{y}_2 & & s^3 + s^2 - s + 2 \end{pmatrix}.$$

Since G_L is not a unimodular matrix (the maximum degree of the polynomials on the main diagonal is higher than 0), system (20) is not accessible. It also means that the system can be reduced. That is, one can find the polynomial matrices \tilde{P} and \tilde{Q} that define the reduced linearized system equations $\tilde{P}(s)dy + \tilde{Q}(s)du = 0$ by solving $P = G_L\tilde{P}$ and $Q = G_L\tilde{Q}$ for \tilde{P} and \tilde{Q} , respectively⁴:

$$\tilde{P} = \begin{pmatrix} \frac{1}{y_2}s^2 + \frac{1}{y_2}s & \frac{1}{y_2}s^2 + \frac{1-u_1-u_2}{y_2}s + \frac{1-u_2-\dot{u}_1-\dot{u}_2}{y_2} \\ s & s - u_1 - u_2 + 1 \end{pmatrix},$$

$$\tilde{Q} = \begin{pmatrix} -s - \frac{\dot{y}_2}{y_2} & -s - 1 - \frac{\dot{y}_2}{y_2} \\ -y_2 & -y_2 \end{pmatrix}.$$

In the reduction process we have assumed that $y_2 \neq 0$, (i.e. $y_2 \in \mathcal{S}$). This guarantees that the coefficients of the polynomials are from the field \mathcal{K} . Now, by (17), the one-forms ω_1 and ω_2 may be found, such that $G_L(s)[\omega_1, \omega_2]^T = 0$:

$$\begin{aligned} \omega_1 &= \frac{1}{y_2}d\dot{y}_1 + \frac{1}{y_2}d\ddot{y}_1 + \frac{1 - u_2 - \dot{u}_1 - \dot{u}_2}{y_2}dy_2 + \frac{1 - u_1 - u_2}{y_2}d\dot{y}_2 \\ &\quad + \frac{1}{y_2}\dot{y}_2 - \frac{\dot{y}_2}{y_2}du_1 - d\dot{u}_1 - \left(1 + \frac{\dot{y}_2}{y_2}\right)du_2 - d\dot{u}_2, \\ \omega_2 &= d\dot{y}_1 + (1 - u_1 - u_2)dy_2 + d\dot{y}_2 - y_2du_1 - y_2du_2. \end{aligned} \tag{21}$$

Note that the one-form ω_2 is exact, $\omega_2 = d\phi_{r2}$, but ω_1 is only closed, and one has to multiply it by an integrating factor y_2 to obtain the exact one-form $d\phi_{r1} = y_2\omega_1$. Altogether, it means multiplying ω by the unimodular matrix

$$U = \begin{pmatrix} y_2 & 0 \\ 0 & 1 \end{pmatrix}$$

from the left, which can be considered as choice of the different gcd⁵

$$G'_L = G_LU^{-1} = \begin{pmatrix} 1 & 0 \\ -s^2 + 1 & s^3 + s^2 - s + 2 \end{pmatrix}.$$

Indeed, the system $P(s)dy + Q(s)du = 0$ may be rewritten as

$$G_L(s)U^{-1}U \begin{pmatrix} \omega_1 \\ \omega_2 \end{pmatrix} = G'_L(s)U \begin{pmatrix} \omega_1 \\ \omega_2 \end{pmatrix} = G'_L(s) \begin{pmatrix} d\phi_{r1} \\ d\phi_{r2} \end{pmatrix} = 0.$$

⁴ For the sake of simplicity the polynomial coefficients (elements of \mathcal{K}) are written in the form $\frac{a}{b} = a^{-1}b$. No confusion is caused since only left fractions are considered.

⁵ Recall that the gcd of P and Q is not unique.

The reduced i/o equations can be found by integrating $d\varphi_{r1}$ and $d\varphi_{r2}$:

$$\begin{aligned} y_2 - u_2 y_2 - y_2 \dot{u}_1 - y_2 \dot{u}_2 + \dot{y}_1 + \dot{y}_2 - u_1 \dot{y}_2 - u_2 \dot{y}_2 + \ddot{y}_1 + \ddot{y}_2 &= 0, \\ y_2 - u_1 y_2 - u_2 y_2 + \dot{y}_1 + \dot{y}_2 &= 0. \end{aligned} \quad (22)$$

Note that system (20) can be rewritten as $F_1(\cdot) = \varphi_{r1} = 0$, $F_2(\cdot) = \varphi_{r1} + 2\varphi_{r2} - \dot{\varphi}_{r2} - \ddot{\varphi}_{r1} + \ddot{\varphi}_{r2} + \ddot{\varphi}_{r1} = 0$, but the reduced Eq. (22) are not in the form (1). It is easy to notice that Eq. (22) can be simplified further and transformed into the form (1), using the linear i/o equivalence transformations [11]. Namely, choosing $\bar{\varphi}_{r1} := \varphi_{r1} - \dot{\varphi}_{r2} = y_2 - u_2 y_2 + \dot{y}_1$ and $\bar{\varphi}_{r2} := \bar{\varphi}_{r1} - \varphi_{r2} = u_1 y_2 - \dot{y}_2$ yields the i/o equivalent system description

$$\dot{y}_1 = u_2 y_2 - y_2, \quad \dot{y}_2 = u_1 y_2. \quad (23)$$

Replacing (22) by (23) may be just interpreted as a different choice of the gclid

$$\bar{G}_L = \begin{pmatrix} s+1 & s \\ 3 & s^2+2 \end{pmatrix}.$$

Really, note that matrices G'_L and \bar{G}_L are column equivalent due to the relation

$$\bar{G}_L = G'_L V, \quad \text{where } V = \begin{pmatrix} s+1 & s \\ 1 & 1 \end{pmatrix}$$

is a unimodular matrix since

$$V^{-1} = \begin{pmatrix} 1 & -s \\ -1 & s+1 \end{pmatrix}.$$

Therefore

$$G_L(s) \left[\tilde{P}(s) dy - \tilde{Q}(s) du \right] = \bar{G}_L(s) \left[\bar{P}(s) dy - \bar{Q}(s) du \right] = \bar{G}_L(s) \left[\begin{pmatrix} s & 1-u_2 \\ 0 & s-u_1 \end{pmatrix} dy - \begin{pmatrix} 0 & -y_2 \\ -y_2 & 0 \end{pmatrix} du \right].$$

Now the one-forms, corresponding to the irreducible equations, are simpler

$$\bar{P}(s) dy - \bar{Q}(s) du = \begin{pmatrix} dy_1 + (1-u_2)dy_2 - y_2 du_2 \\ dy_2 - u_1 dy_2 - y_2 du_1 \end{pmatrix},$$

and are both exact. Their integration will yield the irreducible i/o equations (23). According to Definition 14, systems (20) and (23) are transfer equivalent. Finally, note that (20) can be rewritten as $F_1(\cdot) = \bar{\varphi}_{r1} + \ddot{\varphi}_{r1} + \ddot{\varphi}_{r2} = 0$, $F_2(\cdot) = 3\bar{\varphi}_{r1} - \ddot{\varphi}_{r2} + 2\bar{\varphi}_{r2} = 0$.

As recalled in the introduction, reduction of the i/o equations is an integral part of finding the minimal realization. The state-space description

$$\dot{x} = f(x, u), \quad y = h(x), \quad x(t) \in \mathbb{R}^n \quad (24)$$

is said to be a *realization* of the set of the i/o equations (1) if equations (1) and (24) have the same solution sets $\{u(t), y(t)\}$. System (1) is called *realizable* if the state-space form (24) exists for it. The realizability property can be checked using the sequence of subspaces (4): system (1) has an observable realization in the form (24) iff the subspace \mathcal{H}_{r+2} , r being the highest order of input derivative in (1), is integrable. Integrating the basis vectors of \mathcal{H}_{r+2} yields the state coordinates [3]. Moreover, note also that the realization of the i/o equations (1) is accessible, iff Eq. (1) are irreducible [6].

Observe that the realization of the reducible Eq. (20), as expected, is not accessible. Indeed, the state variables for system (20) may be defined as $x_1 = y_1$, $x_2 = y_2$, $x_3 = \dot{y}_2 - u_1 y_2$, $x_4 = \dot{y}_2 - y_2 \dot{u}_1 - u_1 \dot{y}_2$, $x_5 = \dot{y}_1 - u_2 y_2$, which yield the state equations $\dot{x}_1 = u_2 x_2 + x_5$, $\dot{x}_2 = u_1 x_2 + x_3$, $\dot{x}_3 = x_4$, $\dot{x}_4 = -3x_2 - 2x_3 - 3x_5$, $\dot{x}_5 = -x_2 - u_1 x_2 - x_3 - x_4 - x_5$. According to [6], the system is accessible iff $\mathcal{H}_\infty = \{0\}$. For the state equations above we obtain $\mathcal{H}_\infty = \text{span}_{\mathcal{X}} \{dx_3, dx_4, dx_2 + dx_5\} \neq \{0\}$.

For Eq. (23) one may choose the state variables as $x_1 = y_1$ and $x_2 = y_2$, and then the state equations have the form $\dot{x}_1 = (u_2 - 1)x_2$, $\dot{x}_2 = u_1 x_2$. For the latter system $\mathcal{H}_\infty = \{0\}$, thus the accessibility condition is fulfilled.

The i/o equation in Example 3 below lacks the state-space form.

Example 3. Consider the model of “ball and beam”, [6,19], described by the SISO equation

$$\left(\frac{J}{R^2} + m\right)\ddot{y} + mg \sin u - mru^2 = 0, \quad (25)$$

where angle u is the input of the system, y is the position of the ball and is considered as the output of the system, J , R , m , and g are constant parameters.

Compute $\mathcal{H}_1 = \text{span}_{\mathcal{X}}\{\text{d}y, \text{d}\dot{y}, \text{d}u, \text{d}\dot{u}\}$, $\mathcal{H}_2 = \text{span}_{\mathcal{X}}\{\text{d}y, \text{d}\dot{y}, \text{d}u\}$, and $\mathcal{H}_{r+2} = \mathcal{H}_3 = \text{span}_{\mathcal{X}}\{\text{d}y, A\text{d}y - 2mR^2y\dot{u}\text{d}u\}$, where $A = J + mR^2$. The subspace \mathcal{H}_3 is not integrable and thus the i/o equation (25) is not realizable. However, the accessibility of (25) may be checked directly by Theorem 11. Globally linearized equations (10) for system (25) are represented by matrices

$$P = \left(-\left(m + \frac{J}{R^2}\right)s^2 + m\dot{u}^2\right), \quad Q = (2my\dot{u}s - gm\cos u).$$

Since P and Q both contain a single element, the transformation of the composite matrix $[P|Q]$ into the triangular form reduces to the computation of the gcd of two polynomials. We obtain $G_L = (\gamma) = (-[Ag^2 \cos u^2 + 8Ay^2\dot{u}^2 + 2gy((3A - 2J)\dot{u}^2 \sin u + 3A\ddot{u}\cos u) + 2Ay\dot{u}(3g\cos u + 4y\ddot{u}) + y^2(8(J - A)\dot{u}^4 + 8A\ddot{u}^2 - 4A\ddot{u}\dot{u})]/(2Ry\dot{u})^2)$. For $\dot{u} \in \mathcal{S}$, we assume that $\dot{u} \neq 0$. Since $\deg \gamma = 0$, the matrix G_L is unimodular and thus, system (25) is, by Theorem 11, accessible.

Example 4. In the paper it was silently assumed that $F(0, \dots, 0) = 0$ in (12), yielding $\varphi_r = 0$. This assumption is not natural in the nonlinear case, though widely used in the literature (for instance in [6]). However, in some (rare) cases such assumption may yield a situation when the *irreducible one-form* for the system can be found, while the *reduced i/o equations* do not exist [26]. Such an example is given by the i/o equation

$$\varphi := \ddot{y}u - \dot{y}\dot{u} = 0. \quad (26)$$

In [26] it was shown that the minimal realization⁶ of (26) is not accessible. The irreducible form of (26) is $\varphi_r = \dot{y}/u$, but taking $\varphi_r = 0$ results in a degenerate i/o equation $\dot{y} = 0$. For this reason system (26) is considered irreducible in [26] (and its realization minimal), and it serves as a counterexample for the fact that the minimal realization is accessible. This conclusion is in disagreement with the classical realization theory. Replacing the assumption $F(0, \dots, 0) = 0$ by $F(c, 0, \dots, 0)$, as done in [2], allows us to avoid such mismatch. However, understanding c as a variable leads to several irreducible equations (one for each fixed c). Establishing a theory without the restriction $F(0, \dots, 0) = 0$ brings along unexpected results. Namely, a time-invariant reducible i/o equation of the form (1) may admit only a time-varying irreducible equation; thus extension is neither trivial nor direct (see more in [15]).

Computing the transfer function for system (26) yields: $p := p_{11} = us^2 - \dot{u}s$, $q := q_{11} = \dot{y}s - \ddot{y}$. Transfer function $f := p^{-1}q = (us^2 - \dot{u}s)^{-1}(\dot{y}s - \ddot{y})$. Since p and q have the gcd $g_L := g_{11} = u^2s$ with the degree 1 and from $p = g_L\tilde{p}$, $q = g_L\tilde{q}$, we obtain $\tilde{p} = (1/u)s$ and $\tilde{q} = (\dot{y}/u^2)$. So, the transfer function may be reduced: $f = \tilde{p}^{-1}g_L^{-1}g_L\tilde{q} = \tilde{p}^{-1}\tilde{q} = s^{-1}(\dot{y}/u)$. It is obvious that $\dot{y}/u = 0$ is a singular point of the transfer function f . Moreover, since $\varphi = kF(\varphi_1, \dot{\varphi}_r) = u^2\dot{\varphi}_r$, one has to assure that $u \neq 0$.

6. CONCLUSIONS

The paper presented accessibility definition and a computation-oriented necessary and sufficient accessibility condition for nonlinear multi-input multi-output differential equations which are formulated in terms of the gcd of the polynomial matrices associated with the globally linearized system equations.

⁶ In the sense of state dimension.

Note that the polynomials are defined over the differential field, and unlike the linear case, belong to a non-commutative polynomial ring. On the basis of the above condition a constructive algorithm using the Euclidean left division is suggested to examine the system accessibility. The proposed condition and algorithm are consistent with those for the linear system and SISO nonlinear system in [27]. Note that though the accessibility property can be checked within the polynomial approach, this is not so with the system reduction. The algorithm described in this paper results in the vector of one-forms $\omega = [\omega_1, \dots, \omega_p]^T$, related to the irreducible system description. To find the irreducible equations themselves, one has to integrate ω . Though the set of one-forms $\omega_1, \dots, \omega_p$ is, in principle, proved to be integrable, in general one has to multiply ω with the unimodular polynomial matrix from the left to make the components of the one-forms exact. This amounts to finding another gcd from the class of all column-equivalent gcds.

Recall that the reduction algorithm, based largely on the proof of Theorem 7, does not necessarily result in the reduced equations of the form (1). Though there exist the algorithm to transform the reduced equations into such a form [11], it would be nice to develop the reduction algorithm that will give directly the reduced equations in the form (1).

ACKNOWLEDGEMENTS

The work of Ü. Kotta and M. Tõnso was supported by the European Union through the European Regional Development Fund and the target funding project SF0140018s08 of the Estonian Ministry of Education and Research. M. Tõnso was additionally supported by the Estonian Science Foundation grant No. 8787. The work of Y. Kawano was done during his visit to the Institute of Cybernetics at Tallinn University of Technology and was supported by the Development of International Network for Training of Young Researchers Exploring Multidisciplinary Fields No. J091113016 from the Japan Society for the Promotion of Science. The first two authors thank Dr Claude Moog for the inspiring discussion.

REFERENCES

1. Aranda-Bricaire, E., Moog, C. H., and Pomet, J.-B. A linear algebraic framework for dynamic feedback linearization. *IEEE Trans. Autom. Control*, 1995, **40**(1), 127–132.
2. Bartosiewicz, Z., Kotta, Ü., Pawluszewicz, E., Tõnso, M., and Wyrwas, M. Reducibility condition for nonlinear discrete-time systems: behavioral approach. *Control and Cybernetics*, 2013, **42**(2), 329–347.
3. Belikov, J., Kotta, Ü., and Tõnso, M. Adjoint polynomial formulas for nonlinear state-space realization. *IEEE Trans. Autom. Control*, 2014, **59**(1).
4. Bell, D. J. and Lu, X. Y. Differential algebraic control theory. *IMA J. Math. Control Info.*, 1992, **9**(4), 361–383.
5. Bronstein, M. and Petkovšek, M. An introduction to pseudo-linear algebra. *Theor. Comput. Sci.*, 1996, **157**(1), 3–33.
6. Conte, G., Moog, C. H., and Perdon, A. *Algebraic Methods for Nonlinear Control Systems*. Springer-Verlag, London, 2007.
7. Fu, G., Halás, M., Kotta, Ü., and Li, Z. Some remarks on Kähler differentials and ordinary differentials in nonlinear control theory. *Syst. Control Lett.*, 2011, **60**(9), 699–703.
8. Halás, M. An algebraic framework generalizing the concept of transfer functions to nonlinear systems. *Automatica*, 2008, **44**(5), 1181–1190.
9. Halás, M. and Kotta, Ü. Transfer functions of discrete-time nonlinear control systems. *Proc. Estonian Acad. Sci. Phys. Math.*, 2007, **56**(4), 322–335.
10. Kolchin, E. R. *Differential Algebra and Algebraic Groups*. Academic Press, 1973.
11. Kotta, Ü., Bartosiewicz, Z., Nõmm, S., and Pawluszewicz, E. Linear input-output equivalence and row reducedness of discrete-time nonlinear systems. *IEEE Trans. Autom. Control*, 2011, **56**(6), 1421–1426.
12. Kotta, Ü., Bartosiewicz, Z., Pawluszewicz, E., and Wyrwas, M. Irreducibility, reduction and transfer equivalence of nonlinear input-output equations on homogeneous time scales. *Syst. Control Lett.*, 2009, **58**(9), 646–651.
13. Kotta, Ü., Kotta, P., Nõmm, S., and Tõnso, M. Irreducibility conditions for continuous-time multi-input multi-output nonlinear systems. In *2006 9th International Conference on Control, Automation, Robotics and Vision*. Singapore, 2006, 807–811.
14. Kotta, Ü., Leibak, A., and Halás, M. Non-commutative determinants in nonlinear control theory: preliminary ideas. In *International Conference on Control, Automation, Robotics & Vision*. Hanoi, Vietnam, 2008, 815–820.
15. Kotta, Ü., Moog, C. H., and Tõnso, M. The minimal time-varying realization of a nonlinear time-invariant system. In *Proceedings of the 9th IFAC Symposium on Nonlinear Control Systems (NOLCOS)*. France, Toulouse, 2013, 518–523.

16. Kotta, Ü. and Tõnso, M. Realization of discrete-time nonlinear input-output equations: polynomial approach. *Automatica*, 2012, **48**(2), 255–262.
17. Lu, X. Y. and Bell, D. J. Minimal equivalence of differential io systems. *IMA J. Math. Control Info.*, 1999, **16**(2), 165–178.
18. McConnell, J. C. and Robson, J. C. *Noncommutative Noetherian Rings*. Birkhäuser, 1987.
19. Messner, W. and Tilbury, D. Control tutorials for MATLAB and simulink, 2011. <http://ctms.engin.umich.edu/CTMS/index.php?example=BallBeam§ion=SystemModeling> [accessed 16.10.2013].
20. Ondera, M. The use of Maple in computation of generalized transfer functions for nonlinear systems. In *Innovative Algorithms and Techniques in Automation, Industrial Electronics and Telecommunications*. Springer, Netherlands, 2007, 275–280.
21. Pommaret, J.-F. *Partial Differential Control Theory. Vol. II Control Systems*. Mathematics and its Applications, Vol. 530, Kluwer Academic Publishers, Dordrecht, 2001.
22. Popov, V. M. Some properties of the control systems with irreducible matrix-transfer functions. *Lecture Notes Math.*, 1969, **144**, 169–180.
23. Sontag, E. From linear to nonlinear: some complexity comparisons. In *IEEE Conference on Decision and Control*. New Orleans, 1995, 2916–2920.
24. Van der Schaft, A. J. On realization of nonlinear systems described by higher order differential equations. *Math. Syst. Theory*, 1987, **19**(1), 239–275.
25. Van der Schaft, A. J. Transformations of nonlinear systems under external equivalence. In *New Trends in Nonlinear Control Theory, LNCIS*, **122**, 33–43. Springer-Verlag, Berlin, Germany, 1989.
26. Zhang, J., Moog, C. H., and Xia, X. Realization of multivariable nonlinear systems via the approaches of differential forms and differential algebra. *Kybernetika*, 2010, **46**(5), 799–830.
27. Zheng, Y., Willems, J. C., and Zhang, C. A polynomial approach to nonlinear system controllability. *IEEE Trans. Autom. Control*, 2001, **46**, 1782–1788.

Mittelineaarse juhtimissüsteemi ligipäasetavuse tingimus polünoomide terminites

Ülle Kotta, Maris Tõnso ja Yu Kawano

Lineaarse süsteemi juhitavuse mõistet võib mittelineaarsete süsteemide jaoks üldistada mitmel erineval viisil, käesolevas artiklis on uuritud üht võimalikku üldistust, nimelt süsteemi ligipäasetavust. On esitatud tarvilik ja piisav tingimus mittelineaarse süsteemi ligipäasetavuse kontrollimiseks, kusjuures süsteem on kirjeldatud sisendeid ning väljundeid siduvate kõrgemat järku diferentsiaalvõrrandite abil. Täisdiferentsiaali operaatorit kasutades on süsteemi võrrandid lineariseeritud globaalselt ja seejärel esitatud kahe maatriksi abil, mille elemendid on lineaarse süsteemi korral polünoomid üle reaalarvude korpuse, kuid mittelineaarse süsteemi korral (mittekommutatiivsed) polünoomid üle meromorfsete funktsioonide korpuse. Süsteem on ligipäasetav parajasti siis, kui maatriksite suurim vasakpoolne ühistegur on unimodulaarne, st selle pöördmaatriks on samuti polünoommaatriks. Esitatud tingimus lubab kontrollida ka selliste süsteemide ligipäasetavust, mis ei ole kirjeldatavad olekuvõrrandite abil. Suurima vasakpoolse ühisteguri abil saab leida mittelineaarse süsteemi ligipäasetava alamsüsteemi, mis on sobiv lähtepunkt jälgitavate ja ligipäasetavate olekuvõrrandite leidmiseks. Ühtlasi võimaldab tingimus kontrollida kahe mittelineaarse süsteemi ekvivalentsust, kusjuures ekvivalentsetena mõistame süsteeme, mille ülekandemaatriksid on võrdsed.